User Manual for

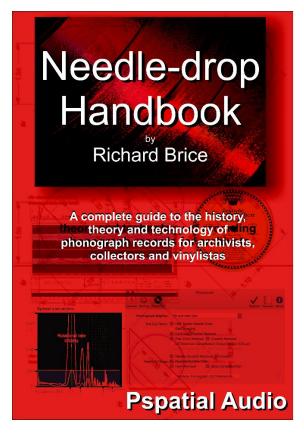
### PHLUX-AI ACTIVE INTERFACE PHONO PREAMPLIFIER



Version 3.2 – PHLUX-AI







The ideal adjunct to the **PHLUX-AI** preamplifier manual is Pspatial Audio's **Needle-drop Handbook** because so much background information is available in this book.<sup>1</sup>

Although the **Needle-drop Handbook** is principally concerned with digital reformatting to preserve the *essence* of audio recordings divorced from their physical carrier, it is as relevant to the audiophile vinylista as it is to the archivist. Both need first-class hardware – where *the rock meets the record*, analogue discs are ultimately a precision mechanical medium and demand precision mechanisms to replay them.

The first five chapters of the **Needle-drop Handbook** are devoted to the mechanical issues of playing records: the mechanics of constant-speed; of tracking and tracing distortion; and of cartridge tracking and groove geometry.

The equipment to cut the master acetates is covered in chapter six of the **Needle-drop Handbook**, as are the subsequent electroforming and compression-moulding processes required to produce commercial records. Only an understanding of how records are made, alerts us to the problems which arise in manufacturing and to the extra information which is encoded in the *dead-wax*.

The care of records – when playing them and storing them - is covered in chapter seven along with information of the care and maintenance of the replay stylus.

The signal sent to the cutter-chisel when cutting a record is equalised, whereby the bass is cut and the treble boosted. The equalisation of modern records is described in chapter eight of the **Needle-drop Handbook** and the equalisation of pre-stereo records has a chapter all to itself in chapter ten.

The revised third edition of the **Needle-drop Handbook** has a series of afterword chapters on: noise mechanisms, both mechanical and electronic; lubrication; filter design; and advice on loudspeakers and amplifiers. Before we get there, there are a series of chapters looking at more unusual forms of records of interest the collector: shellac discs; quadraphonic discs; cylinder records; soundtrack discs for early "talking pictures"; radio transcription discs; and novelty records. We also devote a chapter (thirteen) to looking at popular upgrades and tweaks to turntables and ancillary equipment to see if they work or not.

**Richard Brice** France 2025

<sup>&</sup>lt;sup>1</sup> Go to <u>https://pspatialaudio.com/NDH\_book2.htm</u> to purchase a copy of the **Needle-drop Handbook**.



### Chapter 1 Introduction

#### Why buy a Phædrus Audio phono preamplifier?



Phædrus Audio's **PHLUX-AI** preamplifier is designed for all vinylistas – both the hard-line analogue lover and the modernist. The **PHLUX-AI** preamplifier may be used as a no-compromise, high-quality stand-alone phono preamplifier, or in conjunction with Pspatial Audio's **Stereo Lab** app' which runs under OS X on the Apple Mac.

For the analogue purist, the **PHLUX-AI** preamplifier is designed to offer state-of-the -art, low-noise preamplification in a unique, and innovative

bridge-based preamplifier configuration. It also employs circuit techniques from the summits of analogue design (for example, the legendary WV2 preamplifier from Neumann).

For the modernist (and the archivist), using the **Stereo Lab** app' isn't just technology for technology's sake. To the purist, it may seem sacrilege to digitise the audio from a record in a, so called, *needle-drop* recording. But digitisation is a necessary stage in applying the most powerful technology of our age – modern computing – to recovering the best possible signal quality from a record. We don't cavil when modern signal processing is used to reconstruct lost architecture or reveal the radiance of a painting. In our view, it's churlish to gainsay the benefits of this technology to phonograph records. For example, phase-linear warp and rumble filtering is only possible in the digital domain and reveals a bottom octave with unprecedented clarity and realism. The accuracy of RIAA equalisation and perfect channel balance ensures an uncoloured sound with superb stereo sound staging. And the flexibility in recording characteristics and intelligent noise-reduction will bring to life records you thought mediocre. Moreover, **Stereo Lab** also includes software decoders for CD-4, *QS* and *SQ* quadraphonic and *Ambisonic* LPs; thereby bringing life to forgotten, analogue, multichannel gems.

For **Stereo Lab** to process needle-drop captures and equalise them, it needs recorded files of the "raw" signal, direct from the groove, unfettered by analogue equalisation. And that is where the **PHLUX-AI** preamplifier comes in. It is a phono preamplifier of the very highest quality but also provides a record output *without* equalisation, so that the signal fed to the computer is the closest version of the signal direct from the groove. The **PHLUX-AI** preamplifier has various options to augment this basic task to aid with monitoring and to ease integrating these units into a high-quality audio system. These are described in the pages that follow.



### Features

#### PHLUX-AI unit

- Bridge-based preamplifier with unique, low noise circuit architecture
  - Supports a wide range of moving-magnet cartridges
    - Also moving-coil, active, and strain-gauge types with added options
- Line level (non-equalised) output available for direct connection to external soundcard for use with **Stereo Lab** software
- USB powered all internal rails derived from USB V<sub>bus</sub>
  - o Dual, internal high-stability bandgap regulation to eliminate PSU noise

#### Input/output options

#### Moving-coil input options

- Support for modern as well as historical MC cartridges (e.g. Neumann DST)
  - $_{\odot}$  Two gain versions available (500  $\mu$ V/ 150  $\mu$ V sensitivity)
    - Third, OdB version offers a high-Z input (for moving-iron types)
- State of the art low-noise design: -141dBu EIN at highest gain
- Variable load adjustment

#### **PHLUX** input option

• Provides phantom power to Phædrus Audio's **PHLUX** active cartridges

#### Strain-gauge input option

• Semiconductor cartridge support

#### SYNOPSIS soundcard option

- Stand-alone needle-drop solution
- Driverless, class-compliant USB protocol controller
  - Works with all common operating systems
- USB input-output connection to computer

#### Neumann WV2 output option



- Raises output to full professional output level (-6dBu = standard level)
- Based on output stage of Neumann WV2 playback equaliser for the AM-32b lathe
  - $\circ$   $\,$  Considered to be the Holy Grail of RIAA preamplifiers
- Transformer coupled output for massive signal headroom



#### **Processing options**

#### EQ (RIAA) option

• Provides a precision RIAA equalised output

#### FRANCINSTIEN stereo enhancement option

• Adds stereo enhancement to EQ option

#### **PHOENIX** EQ and signal processing

• Adds processing for shellac-era records to EQ option. Selected by push-switch.

#### Sigē option

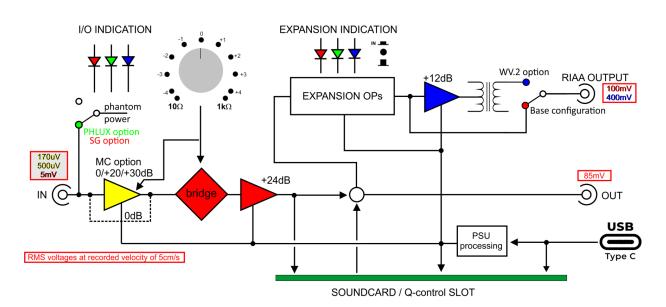
- Power supply upgrade
  - Supercapacitor reservoir

#### **Preamplifier option**

• Volume control and option bundle to provide a complete preamplifier solution

#### **Facilities diagram**

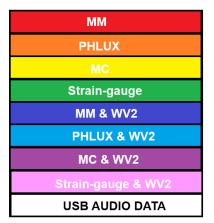
The functionality of equipment like the **PHLUX-AI** preamplifier is often most rapidly grasped with a facilities diagram, like the one given below.



A larger version of this facilities diagram is available in appendix 5.



#### **Multicolour LED configuration indicators**



The left-hand front panel multicolour indicator indicates the input and output (I/O) options fitted in the **PHLUX-AI** preamplifier, according to the table (left).

When a **SYNOPSIS** soundcard is fitted, the LED continues to indicate the input configuration, but when valid audio data is passed over the USB interface, the indicator illuminates WHITE.

Other expansion options: **PHOENIX** shellac processing; **FRANCINSTIEN** stereo correction; etc. are indicated by the right-hand multicolour LED configuration indicator. These are defined right. Note that the WHITE colour signal has different roles according to the options fitted. See elsewhere in the manual for the meaning of the WHITE signal for specific options.



# Chapter 2



Before using any piece of equipment manufactured by Phædrus Audio, be sure carefully to read the applicable items of these operating instructions and the safety suggestions. Keep them for future reference. Follow the warnings indicated on the unit, as well as in these operating instructions.

### THE USER SHOULD NOT ATTEMPT TO SERVICE THE UNIT. ALL SERVICING SHOULD BE REFERRED TO QUALIFIED SERVICE PERSONNEL OR FACTORY ONLY.

Phædrus Audio products should NEVER be in any other way energised when the case is opened and/or the circuit boards are accessible.

#### **General Safety Instructions**

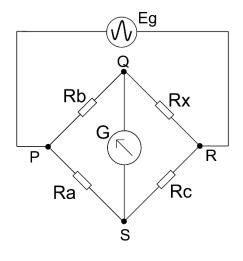
- Do not operate this equipment near any source of water or in excessively moist environments.
- Keep this equipment away from babies, children and pets.
- Do not let objects fall, or liquids be spilled, onto the enclosure.
- Situate this equipment away from heat sources or other equipment that produce heat.
- Ensure this equipment has adequate ventilation.
- When cleaning this equipment, remove all connections to the unit; including power and gently wipe with a clean, lint-free cloth; if necessary, gently moistened with lukewarm or distilled water. Use a dry lint-free cloth to remove any remaining moisture. NEVER use aerosol sprays, solvents, or abrasives on this equipment.
- This equipment should be serviced by qualified service personnel or returned to Phædrus Audio when: an object (or objects) have fallen into the enclosure; or liquid has fallen into, or been spilled into the unit; or the unit has been exposed to rain or high humidity; or the unit does not operate normally or exhibits a marked change in performance; or the unit has been dropped, or the enclosure has been damaged.



Chapter 3

### **Bridge preamplifier theory**

#### Wheatstone's bridge



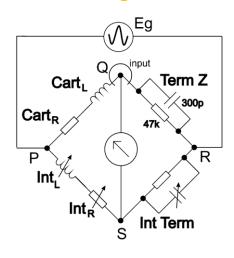
In about 1843, Sir Charles Wheatstone designed a circuit called a *bridge* which gave an accurate method for measuring *resistance*. The circuit is given left. In this circuit, resistors **Ra**, **Rb**, **Rc**, are all known values. **Rx** is the unknown resistance.

The full explanation of this circuit is in almost every textbook on electricity, so we won't repeat that here. The important feature to appreciate is that when the bridge is said to be *balanced*, current through the galvanometer **G** falls to zero, which occurs in the condition,

Ra / Rb = Rc / Rx

In Wheatstone's time, this relation was important because from this it was possible to deduce the resistance of Rx according to the formula,

 $Rx = (Rb / Ra) \times Rc$ 



#### Maxwell's bridge

Maxwell's bridge is a development of the Wheatstone bridge, named after James Clerk Maxwell, who first described it in 1873. Maxwell's bridge extends the use of Wheatstone's bridge to measure reactive (not simply resistive) circuit elements.<sup>2</sup> An example is illustrated. This bridge circuit relates directly to the application in which a moving-magnet electrodynamic phono pickup is driving a conventional high-impedance (47k $\Omega$ ) preamplifier input.

**Cart**<sub>R</sub> and **Cart**<sub>L</sub> represent the resistive and inductive components of the cartridge generator

impedance which together we call **Cart**<sub>z</sub>. The termination impedance (*Term*<sub>z</sub>) is made up of the resistance (typically  $47k\Omega$ ) in shunt with the cable capacitance – here shown as 300pF, a typical figure. Note that point **Q** is the coäxial cable input to the preamplifier. Just like the

<sup>&</sup>lt;sup>2</sup> It is often said that Maxwell's bridge is for the measurement of *inductance* and that was indeed Maxwell's original intention. However, it may be said to be a bridge of general reactance – capacitive as well as inductive.



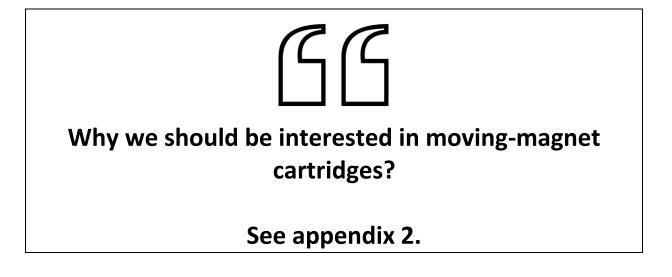
Wheatstone bridge, provided the impedances  $Cart_z$  and  $Int_{Term}$  are balanced with respect to the external impedances, the galvanometer will indicate zero current.

Now, this is of little use practically because point P doesn't exist in reality. The generator voltage Eg is developed across  $Cart_z$  and is not independent. And, even if it was, there would be no way to access point P. Nevertheless, the Maxwell bridge indicates the concept that – provided we balance the impedances of the cartridge and the cable, it ought to be possible to recover the "pure" generator EMF (Eg).

#### **Bridge preamplifier**

In the case of the **PHLUX** active and strain-gauge cartridges, recovering **Eg** is accomplished by the "belt-and-braces" technique of buffering the generator signal, thereby eliminating the effects of the cartridge impedance, the cable, and their mutual interaction. In a similar way, the contributing reactances in the **Cart**<sub>Z</sub> of a moving-coil cartridge are so low that matching the internal impedances is largely unnecessary.

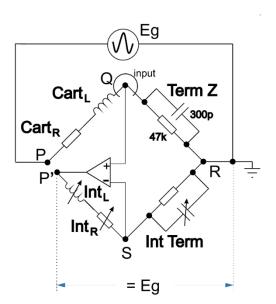
The necessary step in realising a practical preamplifier for moving-magnet cartridges which avoids the need for active electronics in (or very near) the cartridge involves an appreciation that, when the bridge is balanced, point S and point Q are at the same potential. (They must be for there to be no current through the galvanometer.)



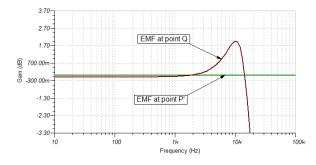


Consider the following circuit in which a high-gain operational amplifier has its two inputs connected to points **Q** and **S**: the non-inverting input fed from point **Q**, and the inverting input from point **S**.

One characteristic of a very high-gain amplifier is that, provided it is not overloaded, the potential between its differential inputs must be very small – because the gain is very high. In this circuit therefore, provided the bridge arms balance – even without a connection to point P – the operational amplifier will force a new point P' to be at the same potential as point P. The original Eg may



thereby be recovered between point **P'** and point **R**. This is the principle behind the bridgebalance preamplifier.



If we arrange *Eg* to be a frequency sweep, we can monitor the voltages versus frequency at point *Q* (the conventional preamplifier input), and at point *P'*. A graph of the results is given. From which, it is very clear how the effect of the interface impedance on the internal impedance of the cartridge is eliminated. Frequency response variation and restriction are both eradicated.

#### **Other innovations**

The bridge arrangement alone does not solve the other issues that the **PHLUX** active cartridge technique addresses. Namely: electronic noise in the high impedance interface between cartridge and preamplifier; and left-right crosstalk in the tiny cables in the tonearm. Likewise, the not inconsiderable problem of calibrating the bridge impedances had to be solved. Various circuit developments were required to address these concerns.



### Chapter 4 Using the PHLUX-AI preamplifier

#### **Connecting the equipment**

The rear panel of the Phædrus Audio PHLUX-AI phono preamplifier is illustrated.



#### Turntable

The leads from the turntable should be connected to the sockets labelled IN. Be sure to respect the channel polarity. The convention used in the Phædrus Audio **PHLUX-AI** is the right channel is the upper of the two connectors. In the base configuration, the sensitivity of the IN connectors on the **PHLUX-AI** preamplifier is appropriate for moving-magnet type cartridges. If the moving-coil option is ordered, the sensitivity is raised to be appropriate for this type of transducer. If the PHLUX option (or the strain-gauge option) is added, phantom power is made available on these connectors.

#### Earth/Ground

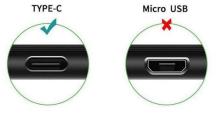
The turntable earth (ground) wire should be connected to the green connector labelled GND. You may need to purchase a (4mm) banana plug to mate with this connector to be sure of making a solid and reliable connection. The socket is a 4mm type and has a tin-plated brass insert designed to accept a wide range of 4mm plugs. Whilst audiophiles love gold-plating, gold-tin is not the ideal electrical contact, so we recommend a 4mm plug of nickel-plated brass.

#### Power Supply (USB)



Finally, connect the USB power. Today, USB power – usually presented via a USB Type-A connector – is ubiquitous. It is available, either via the computer hosting the DAW, or via

stand-alone plug top power supplies, as illustrated. Note that the USB Type-C connector is not the same as the Micro USB connector.





USB power is not known for its stability or lack of noise. To that end, the PHLUX-AI preamplifier incorporates advanced power supply processing which filters any power supply noise and thereby achieves the excellent noise and dynamic range figures it does (see Specifications). You may also like to consider the **Sige** upgrade (see below). The current load on the USB supply due to the PHLUX-AI preamplifier is very light, less than 100mA.<sup>3</sup> The preamplifier may therefore be powered from USB 1.1, 2.0 ports (white plastic inside), or first or second-generation USB 3.0 & 3.1 ports (blue or red plastic inside).



Some customers may wish to use an external, linear power supply type (especially when not using the internal SYNOPSIS soundcard). There are several manufacturers of this type of supply, and they deliver high-quality and low noise 5V DC at very reasonable cost (one is illustrated). An alternative – to keep away from mains power completely – is to use a USB Power bank. If you want to arrange your own PSU, be aware that USB power is specified to be 5V ±5% (4.75V to 5.25V) and these limits must be respected.

**RIAA & OUT** 



Connections to the following soundcard (if used) the configuration of the unit and the options fitted.

Note: there is no iLOOP function as there is on the Groove Sleuth preamplifiers: the middle set of phonos is ALWAYS set to RIAA (as indicated on the rear panel).

#### **User Cases**

In order to see the multiple ways the PHLUX-AI may be employed, we present five user cases: the (Home) Archivist; the Pro Archivist; the Audiophile; the Record Collector; and Mr. Zane Gauge.



<sup>&</sup>lt;sup>3</sup> The internal **SYNOPSIS** soundcard option raises the consumption.



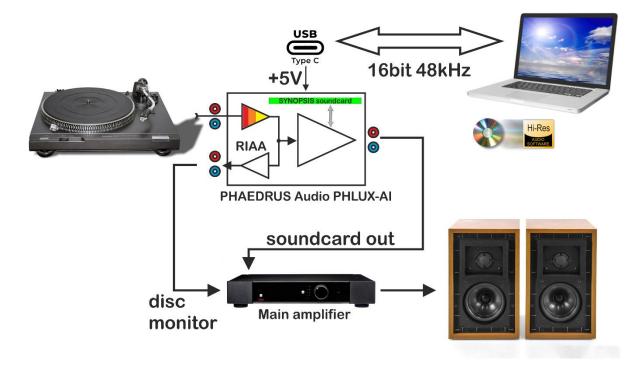
#### User case #1 – the Home Archivist



The **Home Archivist** opted for the most streamlined needle-drop solution, with the internal **SYNOPSIS** soundcard fitted inside the **PHLUX-AI** unit and the RIAA expansion option. She uses the **Hi**-**Res** precision upscaling algorithm in the **Stereo Lab** app' to double the sample-rate of single-rate files and increase the bit depth to 24-bit with a precision dithering algorithm.<sup>4</sup>

The only user input (apart from setting the monitoring volume) is

to select between the RIAA equalised record monitor (on the **RIAA** output connector), and the **SYNOPSIS** soundcard output (on the **OUT** connector). The striped input stage symbol reflects the choice of cartridge options available which include: moving-magnet; movingcoil; strain-gauge; and PHLUX-active. In this case, the **Archivist** opted for a **PHLUX**-active type cartridge. This choice has the added advantage that the Bridge Balance control acts as a fine level control which she uses to "tweak" the record level into the **SYNOPSIS** soundcard.





<sup>4</sup> **Hi-Res** may be selected as a specific process in **Stereo Lab** or it may be enabled as a default processing for all files within the Settings menu.



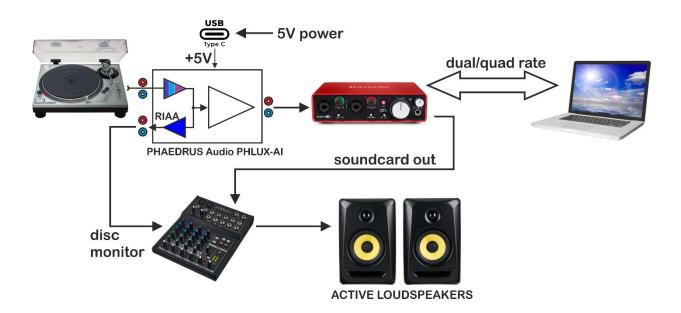
#### User case #2 - the Professional Archivist



#### The Professional Archivist

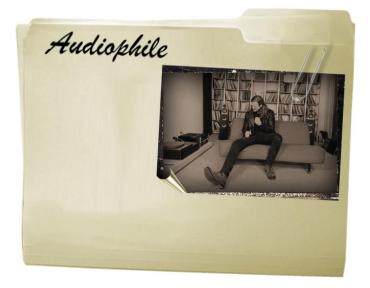
wanted to be able to needledrop record at dual (96kHz) or quad rate (192kHz) sampling, and this required an external soundcard. In order better to match the audio levels from a professional soundcard at the mixer, the Pro Archivist has chosen the **WV2 output option** for the RIAA. The striped input stage symbol reflects the choice of cartridge options available (at time of order) which include: moving-magnet; moving-coil; strain-gauge; and PHLUX-active.

In line with his organisation's policy on pickups for archiving, the **Pro Archivist** chose a classic (Shure) moving-magnet cartridge.



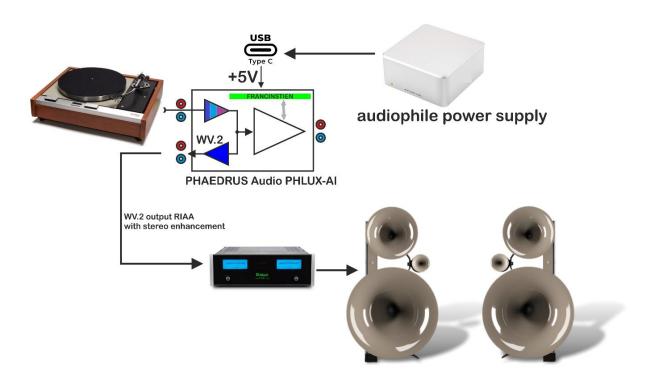


#### User case #3 - the Audiophile



The Audiophile has no interest in a computer connection. Instead, the PHLUX-AI preamplifier is fed from a quiet, audiophile quality USB power supply. He also chose the *Sigē* PSU-upgrade option (see below). The RIAA EQ + FRANCINSTIEN option card is fitted, and the WV2 output option to maximise stereo-imaging and dynamic-range performance. The striped input stage symbol reflects the choice of cartridge options available (at time of order) which include: moving-magnet; moving-

coil; strain-gauge; and **PHLUX**-active. In this case, the **Audiophile** chose an expensive moving-coil type cartridge. The very low moving-coil input stage compliments this cartridge choice perfectly.



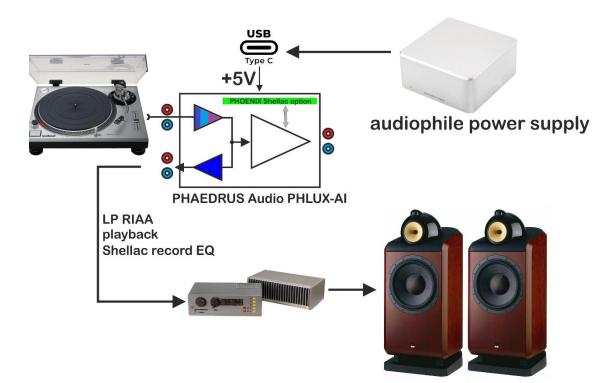


#### User case #4 – the Record Collector



The **Record Collector** has a collection of valuable LPs and shellac records of classic performances. So, she chose the **RIAA + PHOENIX** expansion option (for shellac records) to maximise the quality possible from this ancient medium. The striped input stage symbol reflects the choice of cartridge options available (at time of order) which include: moving-magnet; moving-coil; strain-gauge; and **PHLUX**-active. Our record collector chose a selection of moving-magnet

cartridges, so as to be able to play shellac records, as well as mono and stereo LPs with the appropriate styli.



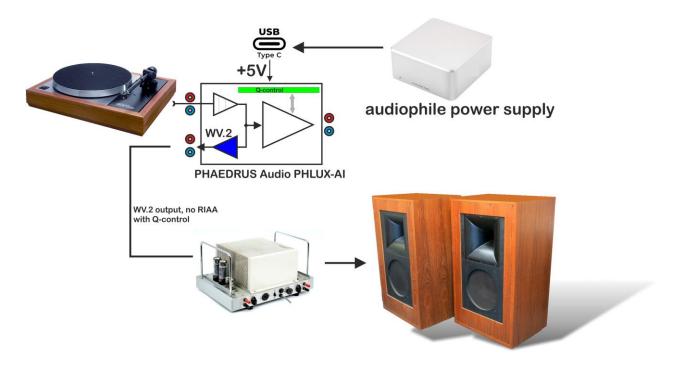


#### User case #5 – Strain gauge cartridge user



Our last user case, **(codename: Mr. Zane Gauge)** chose the historical strain-gauge cartridge input stage option. These unusual, semiconductor cartridges were principally manufactured by Panasonic in the 1970s. Some audiophiles consider these cartridges to offer performance that has never been surpassed. (Mr. Zane Gauge is one such!) When this option is installed in **PHLUX-AI** preamplifier, no EQ is required at the **RIAA** output –

although we recommended the **PHLUX-AI-WV2** output option was purchased. The **OUT** output remains wideband and may be used to decode CD-4 recordings in **Stereo Lab** software (see the **Needle-drop Handbook**).<sup>5</sup>



<sup>&</sup>lt;sup>5</sup> Go to <u>https://pspatialaudio.com/NDH\_book2.htm</u> to purchase a copy of the Needle-drop Handbook.



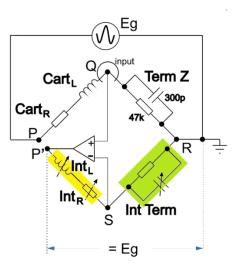
#### **Operating the equipment – the control knob**

The function of the single front-panel control knob varies according to the input option fitted. In the case of the moving-magnet input stage, the control knob adjusts the bridge-balance. This is set various ways as explained below. When the moving-coil input stage is fitted, the control knob adjusts the loading presented to the cartridge, which gives subtle voicing shifts. In the case of the **PHLUX** input option, and when the strain-gauge input option is present, the control is a fine level-control. In the case of the preamplifier option, the control acts as a (full range) volume control.

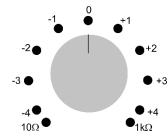
#### Balancing the bridge – moving-magnet cartridges<sup>6</sup>

Referring to Chapter 3, if we look again at the bridge configuration of the **PHLUX-AI** preamplifier, we can clarify that the preamplifier architecture automatically compensates for *Termz* – the lower right limb of the bridge (marked in green). But it is not possible completely to tune automatically the lower left-hand limb (*Cartz*, marked in yellow). Some input is thereby required from the equipment operator.

There are several (4) ways to set the fine bridgebalance control on the front-panel of the **PHLUX-AI** preamplifier. These are detailed below.



It is entirely possible to calibrate the bridge-balance by ear. In fact, this is an excellent approach. It should be noted, this is a *fine balance* adjustment, **the tonal differences are subtle.** There is no element of quality expressed in the control calibration: +4 is no better than -4. The numbers simply relate to the degree of adjustment from the nominal balance condition at the zero (12 o'clock) position.



Set up the equipment and play a disc you know well, monitoring via the internal RIAA or iLOOP outputs. Select the control's central, 12 o'clock, position. If the sound is a little "lean" (bass light), rotate the balance control clockwise until you achieve the best tonal balance. If the sound balance is a little "rich" (bass heavy), rotate the control anticlockwise until you achieve the best tonal balance.

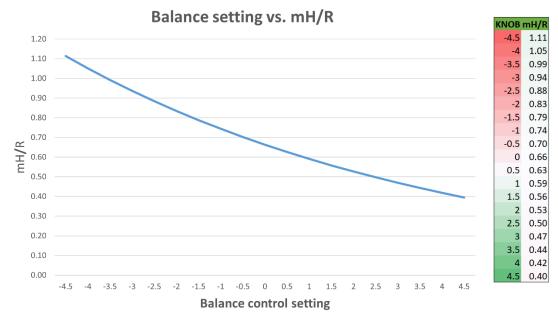
If you feel unsure about setting the bridge-balance by ear, method 2 and 3 represent the engineer's approach.

Consult the manufacturers' information regarding the cartridge coil resistance and inductance. Once you have these figures, substitute the figures in the following equation to derive a value for the bridge factor,  $F_b$ .

<sup>&</sup>lt;sup>6</sup> On early models of the **PHLUX-AI**, the bridge-balance control was simplified to two switches. This restricted the choice of moving-magnet cartridges. The balance control is now a rotary control knob.



Thus, should a cartridge be specified to have a DC resistance of  $800\Omega$  and an inductance of 490mH, **F**<sub>b</sub> will equal 490/800 = 0.61. Select the bridge-balance control position according to the following graph, table, or equation.



Control position = -  $[20.log(1.51 \times F_b)]$ 

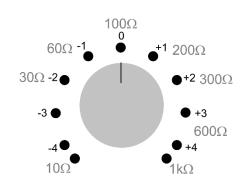
We have not always found that cartridge specifications are available, or entirely reliable – due to manufacturing changes and different approaches to measurement. If you, or an engineer colleague or friend, can measure the coil inductance and resistance of the cartridge with a component bridge, then these values should be substituted in the equation for  $F_b$  and used to select the fine balance setting accordingly. An example of a modern component bridge is illustrated. These are available from online retailers at a very reasonable price.

Finally, ask us at <u>sales@phaedrus-audio.com</u>. We have an extensive cartridge database and can advise.





#### Balancing the bridge – moving-coil cartridges



It is widely believed that input loading adjustment is worthwhile for moving-coil cartridges. To that end, when the **PHLUX-AI** preamplifier has the moving-coil front-end fitted, the bridge-balance knob controls the effect of the input loading on the cartridge by adjusting the mirror internal impedance of the Maxwell bridge.

With the control is in the zero (12 o'clock) position, the cartridge is presented with a  $100\Omega$  load. This is the nominal position. At the counterclockwise

extreme, the effective load falls to  $10\Omega$ , and at the extreme clockwise position, the effective load rises to  $1k\Omega$ , as illustrated. The setting is largely subjective, although some manufacturers suggest a load value for their cartridges.

#### Balancing the bridge – strain-gauge cartridges

When the strain-gauge input option is fitted, there are no reactances to balance within the Maxwell bridge and the control knob acts as a fine level-control (+3.5dB/-6dB).

#### **Balancing the bridge – PHLUX active cartridges**

When the **PHLUX** front-end is fitted in the **PHLUX-AI** preamplifier, the function of the balance knob alters once again. In this case, there are no reactances to balance within the Maxwell bridge and the control knob acts as a fine level-control (+3.5dB/-6dB). This is especially useful if the internal **SYNOPSIS** soundcard is fitted so that digital dynamic range may be optimised.

### Chapter 5

Zweikanal Wiedergabeverstärker WV2

The Neumann WV2 is a vintage Zweikanal-Wiedergabeverstäker (two-channel playback preamplifier). The unit was originally designed for professional vinyl mastering and playback systems. Introduced in 1958, the WV2 was built for demanding, professional use and had a



price tag to match (DM1288 - equivalent to about €5000 today). Today, the mono WV1 and stereo WV2 are considered the *Holy grail* of phono preamplifiers, coveted by enthusiasts seeking the best audio fidelity.

The output stage of the WV2 preamplifier is an E88CC triode driven transformer circuit with an anode load formed of the Bv32 choke. The performance of this stage is closely emulated in the **PHLUX-AI WV2 Output** option.

### Chapter 6

### **PHOENIX option for shellac era records**



Many recording characteristics were used during the *shellac period*, as described in **Needle-drop Handbook**.<sup>7</sup> For serious archive work, an armoury of different equalisation curves and noise-reduction systems is required for the archiving of shellac era discs. Our high-resolution audio processing program **Stereo Lab** provides the appropriate equalisation curves for shellac discs of all eras – all furnished with a number of intelligent noise reduction systems (see dialogue from the app' below).

But what if you just want to listen in real time to a

collection of shellac-era records? That is the rôle of the new **PHOENIX** expansion card for the **PHLUX-AI** preamplifier.

The equalisation curve of the Phædrus Audio **PHOENIX** expansion card was developed with the benefit of Pspatial Audio's years of experience of working with shellac era records. We worked with them and their customers to develop to a compromise equalisation, known as the **Pspatial Audio Universal Shellac** (PAUS) characteristic which serves for most laterally recorded shellac

	None					
	Phonograph EQ A2 (Decca LP)					
	Phonograph EQ A3 (TELDEC/AES)					
	Phonograph EQ B1 (NAB)					
	Phonograph EQ B2 (Columbia LP)					
	Phonograph EQ C (CCIR/Coarse Groove)					
	Phonograph EQ D (Quad 22 all buttons)					
	Phonograph EQ E (500-Flat)					
	Phonograph EQ F (Flat/Acoustic)					
	Phonograph EQ G (European 78s)					
	Phonograph EQ H (American 78s)					
	Phonograph EQ J (Columbia 78s)					
	Phonograph EQ K (Ceramic)					
	Phonograph EQ S (Displacement)					
1	Phonograph EQ A1 (RIAA)					
	Phonograph EQ P1 (Strain-Gauge/Light-Valve Cartridge)					
	Phonograph EQ P2 (DisC Displacement Cartridge)					
	Phonograph anti-EQ A1 (RIAA)					

discs.<sup>8</sup> The **PHOENIX** expansion card also automatically cancels tracing distortion (pinch-effect).

<sup>7</sup> Go to <u>https://pspatialaudio.com/NDH\_book2.htm</u> to purchase a copy of the **Needle-drop Handbook**.

<sup>8</sup> Use **Stereo Lab** software for the reproduction of hill-and-dale discs. See the **Needle-drop Handbook**. Some manufacturers suggest that a flat-response phono preamplifier is the correct solution to replaying 78s. But this is wrong: shellac era records have all required bass boost (with a turnover of about 300Hz) since the late 1920s. Flat-response preamplifiers have their rôle and are essential when using **Stereo Lab** software equalisation. But they are not the best solution to replaying shellac records in real time.



When this option is fitted, **PAUS** equalisation is selected by pressing the front-panel pushswitch to the IN position. The right-hand multicolour LED will indicate WHITE to indicate this EQ is applied.

Chapter 7

### **FRANCINSTIEN** expansion option

The two-channel spatial reproduction system that we know as "stereo" was invented by Alan Blumlein at EMI in suburban London at the turn of the 1930s.<sup>9</sup> Although it took nearly 30 years for Blumlein's ideas to be recognised and brought to the market as a practical stereo system for LP records,<sup>10</sup> Blumlein's ideas were remarkably comprehensive. He even identified a weakness in the stereo system, and invented corrective equipment for it. He called this device the *Stereo Shuffler*. But – almost 100 years after he invented it – most stereo systems remain incomplete with his *Shuffler* largely consigned to history.

#### How stereo works, and what's wrong with it

To understand stereo correction, you must understand a little about three different subjects:

- the physics of sound:
- the psychology of hearing:
- and how recordings are made.

#### A little physics

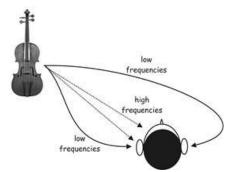
Everyone knows, of course, that sound is waves in air, and that low frequency sounds have long wavelengths and high-frequency sounds have short wavelengths. But most people are surprised by the dimensions of the waves. For example, the lowest note on a bass guitar has a wavelength of 8.5 metres and the wavelength of the highest frequency you can hear is less than 2 cm.

<sup>&</sup>lt;sup>10</sup> Clark, Dutton and Vanderlyn (1958) The "Stereosonic" recording and reproducing system: a two-channel system for domestic tape records JAES Vol. 6, number. 8. April 1958



<sup>&</sup>lt;sup>9</sup> Blumlein, A. (1933) British Patent 394,325 June 14<sup>th</sup>. For a biography of this remarkable man and his untimely death, see <a href="https://en.wikipedia.org/wiki/Alan\_Blumlein">https://en.wikipedia.org/wiki/Alan\_Blumlein</a>

This huge range of wavelengths causes high-frequency sounds to behave differently to low frequency sounds. Essentially, low frequency waves bend around us and aren't absorbed by our bodies. A physicist would say they *diffract*. But high frequency sound waves can't bend around us and our bodies – especially the head – cast an acoustic shadow.



#### A little psychology

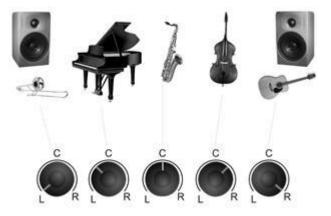
Given this differing behaviour of sounds at low and high frequencies, evolution has provided us with two different processes for determining the direction of sound: one to determine the direction of bass sounds; and another to determine the direction of high-frequency sounds.

#### A little about stereo recording......

The stereo picture is built up in a recording studio by mixing the signals from the microphones and steering them to a place within the stereo image by means of a panoramic potentiometer or *pan-pot*.

The pan-pot positions the contribution from a particular channel in the stereo picture by the ratio of signal it sends to each channel. So, at the extreme anticlockwise position, the entire signal is sent to the left channel, and the contributing microphone signal will appear to come directly from the left loudspeaker. And at the extreme clockwise position, the opposite will be the case, so that the signal will all be sent to the right channel and the microphone signal will all appear to come from the right-hand loudspeaker.

In between there is a progressive relationship so that, as the pan-pot knob is turned clockwise gradually less signal gets sent to the left, and more to the right and the microphone signal appears to move across the stereo field. In the middle of the pan-pot's travel, equal amounts are sent to right and left and the signal appears to come from the centre of the image.<sup>11</sup>



#### The problem

Now, you might be a bit surprised, given that we humans possess two hearing systems, one for high frequencies, and one for low frequencies, that such a simple method of constructing a stereo image with a pan-pot wouldn't provide a perfect illusion. And you would be absolutely right.

<sup>&</sup>lt;sup>11</sup> For the capturing of real sound-fields, Clark et al. (op. cit.) show how the sound-field may be sampled to recreate the appropriate phase-shifts at the listener's ears. There are various techniques. Clark and his team opted for a coincident stereo microphone technique based on crossed figure-of-eight (velocity) microphones.



Essentially a pan-pot steers the low-frequency sounds into one position in the stereo image and steers the high frequencies to another position. The positions are quite close, and that's why stereo has done without correction for 50 years.

The tragedy is that the inventors of the stereo system (Blumlein and the EMI team in the UK) realised this problem when they invented the practical stereo system – nearly a hundred years ago! They even invented a circuit to correct for it. They called it the *Stereo Shuffler*. But their circuit suffered from implementation issues inherent in the technology of the time. It interfered with frequency response and created phase-distortion. So, it was dropped.

And that's meant that Blumlein's stereo system was "broken" – and it has remained that way ever since. Stereophonic sound has never delivered on the promise its inventor had for it nearly a century ago. Rather compromise and "good enough" have ruled the day.

#### **FRANCINSTIEN**

Just before the turn of the century, an alternative *Stereo Shuffler* which implemented the same effect by means of inter-channel crosstalk was proposed.<sup>12</sup> Functionally, the technique was identical to the EMI *Stereo Shuffler* but sidestepped the complications and compromises of EMI's implementation. This is the

system implemented on the expansion card for the **PHLUX-AI** preamplifier.

#### What do I hear?

The result of correctly implemented *Stereo Shuffling* is a transformation in the solidity of the stereo image. High and low frequency components of the instruments "lock-in". Voices and instrument take on a precise position in the image and reverberation takes on a 3D character.



### Chapter 8

### **SYNOPSIS internal soundcard**

The **SYNOPSIS** internal soundcard consists of a stereo audio CODEC and a USB-compliant full-speed protocol controller. The soundcard is class-compliant and does not require drivers – no matter the computer operating system. The **SYNOPSIS** expansion card employs single-rate, 16-bit Delta-Sigma ADCs and DACs. The analogue to digital conversion equipment is calibrated so that 5cm/s RMS lateral groove velocity modulation at 1kHz indicate -18dBFS peak in both the level meters of each stereo channel. This ensures that the full-scale value

<sup>&</sup>lt;sup>12</sup> Brice, R. (1997) Multimedia and Virtual Reality Engineering. Newnes and Brice, R. (1998) Music Engineering. Newnes



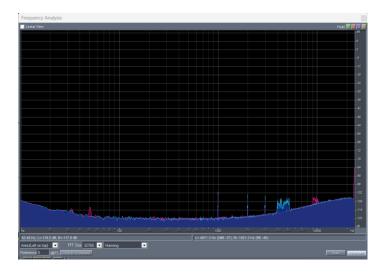
of the digital system is calibrated to -1dB of absolute maximum record velocity (see chapter eight of the **Needle-drop Handbook**).

When this option is fitted, the rôle of the connectors labelled **OUT** changes. Normally these connectors carry the non-equalised signal to an external soundcard. When the **SYNOPSIS** soundcard is fitted, these connectors carry the soundcard output.

#### **Performance considerations**

The ADCs in the **SYNOPSIS** soundcard are 16-bit Delta-Sigma ( $\Delta\Sigma$ ), single-rate types, with a theoretical dynamic range of 16 × 6dB = 96dB.<sup>13</sup> It's interesting to compare the needle-drop capture results using the internal **SYNOPSIS** soundcard with the pure analogue performance measured via our Audio Precision test system and reported in appendix 3 of this manual. The figures reported below are for the **PHLUX-AI** connected directly to a MacBook Pro via a USB interface.

Measured with open circuit input connectors in bridge MM mode with an AC voltmeter, equipped with 300Hz high-pass and (23kHz) band limit filters (see appendix 3), the RMS noise level at the OUTPUT connectors is -73dB relative to maximum output. Captured with the **SYNOPSIS** card, the noise looks like this.



Analysed using *Adobe Audition's* RMS calculator, the RMS noise level is exactly the same, 73dB below maximum modulation (-73dBFS). The **SYNOPSIS** adds no noise to the basic analogue circuit – although there are a couple of low-level signals - these are artefacts from the **SYNOPSIS** ADCs and burst transmissions on the USB 2.0 interface. These look a bit alarming until it is appreciated that these products peak at -102dBFS.

The signal at this point will always be (RIAA) deëmphasised for listening in the **Stereo Lab** app', whereupon the RMS noise measurement falls to -80dBFS. The spectrum is illustrated

 $<sup>^{13}</sup>$  At the core of the Delta-Sigma ( $\Delta\Sigma$ ) ADC lies the paradigm of oversampling, combined with noise shaping and digital filtering. Oversampling involves sampling the analogue input at a significantly higher rate than the Nyquist rate. This disperses quantisation noise across a broader frequency spectrum – ensuring a substantial portion lies beyond the frequency scope of interest. Following digital filtering (decimation), the sample rate is reduced while out-of-band noise is removed.



below. If we filter this with an A-weighting filter, the RMS noise falls to -94dBFS. In the digital domain, this noise is -76dB below standard modulation level, an excellent figure.

With the moving-coil (MC20) option fitted, the audio captured with the **SYNOPSIS** card, and analysed using *Adobe Audition's* RMS calculator, the RMS noise level is 80dB below maximum modulation (-80dBFS). As before, the signal at this point will always be (RIAA) deëmphasised for listening in the **Stereo Lab** app'. Noise on the RIAA equalised signal is - 85dBFS. If we filter this with an A-weighting filter, the RMS noise falls to -95dBFS. In the digital domain, this noise is -77dB below standard modulation level, and nearly 30dB below the vinyl noise-floor.

#### **Single-rate considerations**

The ADCs in **SYNOPSIS** may be set to run at 32, 44.1, 48 kHz sampling rates. This is an uncontroversial limitation if you want to capture and decode CD-4 supersonic carrier records (for which a workable audio bandwidth of 48kHz is required, see the **Needle-drop Handbook**). For this work, you will need an external soundcard capable of capturing files at 96kHz or above.

#### **USB power and standby operation**

The **SYNOPSIS** soundcard – like the rest of the **PHLUX-AI** preamplifier is powered from the five volts present on the USB interface (known as USB  $V_{BUS}$ ). When an attached computer goes into sleep mode, the behaviour of the USB  $V_{BUS}$  depends on a host of design decisions and system settings. These include BIOS/UEFI configuration, and the specific hardware implementation. Refer to your computer manufacturer for information.

The standby power consumption of the **PHLUX-AI** preamplifier is compatible with European Commission's Regulation (EU) 2023/826 with the <u>new rules from 2025</u>, in which power consumption must be less than ½ Watt in all standby configurations. The only circuit option which causes power dissipation to exceed ½ watt is the internal **SYNOPSIS** soundcard. When a valid USB audio computer audio session has terminated and there exists a constant idle state on the USB data bus, the **SYNOPSIS** soundcard enters a suspend state, which reduces overall power consumption of the **PHLUX-AI** to below ½ watt. It remains that way whilst there exists an idle state on the USB data bus. When a new, valid audio session is asserted, the CODEC and protocol controller wake up almost immediately.

### Chapter 9

### **PREVIEW** - Preamplifier option\*

\* This option is not available at launch

The preamplifier option for the **PHLUX-AI** adapts the control knob to operate as a highprecision volume control (of advanced design) and includes a bundle of options to provide a complete control preamplifier. The options included are:

**PHLUX-AI-WV2** - Neumann WV2 output stage (5cm/s velocity  $\rightarrow$  -6dBu)



**EXP-RIAA** – Expansion card to add precision RIAA equalised output **EXP-FRANC** – Stereo "shuffling" enhancement

And either:

PHLUX-AI-MC0 - Optional hi-Z support (5mV sensitivity)
 PHLUX-AI-MC20 - Optional MC support for PHLUX-AI (500μV sensitivity)
 PHLUX-AI-MC30 - Optional MC support for PHLUX-AI (150μV sensitivity)
 PHLUX-AI-PHLUX - Optional PHLUX support
 PHLUX-AI-STRAIN - Optional strain-gauge support

A system drawing of how the preamplifier option is used is given. Note, the **PHLUX-AI** feeds the power amplifier directly.



Note: The SYNOPSIS soundcard is not available in the preamplifier configuration.

## Chapter 10

### Sigē PSU upgrade option

**Sigē** (pronounced *see-gay*) was the personification of silence in the Ancient Greek and Gnostic traditions. In ancient Greek language, **Sigē** represented, not just a lack of noise, but a profound stillness. **Sigē** was not merely the absence of sound: it symbolised an opportunity for prayer and listening. In Gnostic mythology, **Sigē** was the mother of Thought. The **Sigē** option for the **PHLUX-AI** is a supercapacitor-based upgrade to the **PHLUX-AI** frontend, low-noise power supply. **Sigē** employs an electric double layer (EDLC) supercapacitor which removes any noise or interference on the incoming power supply.

An EDLC is a type of supercapacitor that stores energy electrostatically at the interface between the electrodes and an electrolyte. No chemical reaction is needed to store or



release charge, as in a battery: instead, the supercapacitor employs porous carbon electrodes with enormous effective surface area. Instead of a conventional metal oxide dielectric, it uses an electrolyte. At each electrode/electrolyte interface, an electric double layer forms, similar to a capacitor but with nanometre-scale separation ( $\approx 0.3$  to 1 nm). This, and the huge effective surface area of the porous carbon electrodes (thousands of square metres), results in a capacitance measured in farads (F) instead of microfarads ( $\mu$ F).

When the unit is first energised, left-hand front panel multicolour indicator will light white, indicating the presence of the *Sigē* option. During initialisation of the power supply, the audio signals at the RIAA audio outputs will be muted. After about 10 seconds, the white LED will extinguish, and the LED colour will indicate the input option installed. When this change occurs, it indicates that the *Sigē* circuits are active, and the audio circuits will unmute.

Chapter 11

### **Specifications**

#### **PHLUX-AI Preamplifier**

Size: PHLUX-AI: 112 × 50 × 225 mm (W-H-L) Preamplifier Supply: 5V ±5% via USB Type-C connector. Power: Compatible with European Commission's Regulation (EU) 2023/826 with the <u>new</u> <u>rules from 2025</u>. Standby power consumption is less than ½ Watt. Function: Line output to external computer soundcard Sensitivity @ 5cm/s recorded velocity: Nominal 5mV RMS (-44dBu) Input load: Active (Bridge-node) Gain: Nominal 24dB to OUT Signal to noise (rel. 5cm/s velocity, ENBW RIAA): 80dB\* Headroom (rel. 5cm/s velocity): 23dB Crosstalk: Lower than -60dB (1kHz) \*See Appendix 3.

#### **RIAA output**

Function: Adds RIAA-equalised output at middle-pair RIAA connectors Gain (to RIAA output): 26dB at 1kHz Noise figure (all configurations, unweighted in 20kHz): 2dB, this being SNR<sub>out</sub>/SNR<sub>in</sub> Headroom (rel. 5cm/s velocity): 23dB

#### WV2 output

Function: Adds last-stage of Neumann WV2 playback equaliser: raises the level of the RIAAequalised to full operating line-level output (standard level  $\rightarrow$  -6dBu) Gain (to RIAA output): 38dB at 1kHz Noise figure (all configurations, unweighted in 20kHz): 2dB, this being SNR<sub>out</sub>/SNR<sub>in</sub> Headroom (rel. 5cm/s velocity): 23dB



Rumble filter: -1dB@30Hz; -3dB@22Hz; -15dB@10Hz; -40dB@3Hz Maximum output signal level at RIAA output: +18dBu.

#### MC20 Moving coil cartridge option

**Function:** Increases input sensitivity by +20dB **Input load:** Nominal  $100\Omega$  (adjustable with control-knob) **Sensitivity:** Nominal  $500\mu V$  RMS (-64dBu) **Crosstalk:** Lower than -60dB (1kHz) **Signal to noise (rel. 5cm/s velocity, A-weighted in 20kHz):** 70dB to RIAA and MAIN output

#### MC30 Moving coil cartridge option

Function: Increases sensitivity of input by 30dB
Input load: Nominal 100Ω (adjustable with control-knob)
Sensitivity: Nominal 150µV RMS (-74dBu)
Crosstalk: Lower than -60dB (1kHz)
Signal to noise (rel. 5cm/s velocity, A-weighted in 20kHz): 68dB RIAA and MAIN output

#### MC0 Moving coil cartridge option

Function: Provides high-Z input for MI type cartridges: no increase in sensitivity
Input load: 47kΩ nominal
Sensitivity: Nominal 5mV RMS (-44dBu)
Crosstalk: Lower than -60dB (1kHz)
Signal to noise (rel. 5cm/s velocity, A-weighted in 20kHz): 70dB RIAA and MAIN output

#### **PHLUX active cartridge option**

Function: Phantom supply provided at input connector
Input load: >47kΩ nominal
Sensitivity: Nominal 5mV RMS (-44dBu)
Signal to noise (rel. 5cm/s velocity, A-weighted in 20kHz): 70dB RIAA and MAIN output

#### Strain gauge semiconductor cartridge option

**Function:** Phantom supply & LH signal inversion **Input load:** >47kΩ nominal **Sensitivity:** Nominal 5mV RMS (-44dBu) **Signal to noise (rel. 5cm/s velocity, A-weighted in 20kHz):** Depends on cartridge model

#### **SYNOPSIS** soundcard option

Function: Stereo audio CODEC and USB-compliant full-speed protocol controller Drivers: Class-compliant, irrespective of the computer operating system. The SYNOPSIS USB ADC/DAC technology: Single-rate, 16-bit Delta-Sigma (ΔΣ) ADCs and DACs Calibration: 5cm/s RMS lateral groove velocity indicate -18dBFS peak (stereo) DAC resolution: 93dB ADC resolution: 89dB

Phædrus Audio reserves the right to change specifications without notice.



Chapter 12

### Warranty and service

PLEASE RETAIN YOUR SALES RECEIPT. IT IS YOUR PROOF OF PURCHASE COVERING YOUR LIMITED WARRANTY. LIMITED WARRANTY IS VOID WITHOUT SUCH PROOF OF PURCHASE.

#### **Phædrus Audio's Limited Warranty**

Warranty service conditions are subject to change without notice. For the latest warranty terms and conditions and additional information regarding Phædrus Audio limited warranty, please see complete details online at <u>www.phaedrus-audio.com</u>.



#### Appendix 1 - Model Codes/SKUs

The model/order codes for PHLUX-AI preamplifier are: PHLUX-AI - Active interface phono preamp for moving-magnet cartridges PHLUX-AI-MCO - Optional hi-Z support (5mV sensitivity) PHLUX-AI-MC2O - Optional MC support for PHLUX-AI (500μV sensitivity) PHLUX-AI-MC3O - Optional MC support for PHLUX-AI (150μV sensitivity) PHLUX-AI-PHLUX - Optional PHLUX support PHLUX-AI-PHLUX - Optional strain-gauge support EXP-SYNOPSIS – Optional internal soundcard EXP-Sigē-PSU – Optional PSU upgrade (occupies same slot as SYNOPSIS) EXP-RIAA – Optional expansion card to add precision RIAA equalised output PHLUX-AI-WV2 – Optional Neumann WV2 output stage (5cm/s velocity → -6dBu) EXP-PHOENIX – Expansion EQ replay of shellac era records (requires EXP-RIAA option) EXP-FRANC – Optional stereo "shuffling" enhancement (requires EXP-RIAA option) PHLUX-AI-PRE – Volume control and option bundle for complete preamplifier solution

# Appendix 2 – Why moving-magnet cartridges continue to be relevant

Most vinyl-record enthusiasts share a belief – enthusiastically endorsed by cartridge manufacturers – that the more expensive, and more profitable, moving-coil type cartridges outperform the moving-magnet type. This is a simplification bordering on misrepresentation. Market competition was so fierce during the period when records were a mass-market commodity that all the margin was squeezed out of the more popular moving-magnet type cartridge leaving the moving-coil models as the only product in which manufacturers could invest and hope for a reasonable return. History forced the moving-magnet to dominate the lower-end of the cartridge market leaving moving-coils to remain unchallenged at the top-end.



It was not always so. In the 1950s, ceramic (crystal) cartridges were the norm in the home and moving-coil cartridges were standard in broadcasting and in other high-quality applications. It was considered a very great step forward when the American cartridge manufacturer Shure Brothers introduced the first, modern moving-magnet cartridge.<sup>14</sup> As Ben Bauer explains in a paper<sup>15</sup> describing the development of the phono cartridge,

...By devising a moving-magnet pickup, in which the traditional roles of armature and magnet were reversed, it became possible to lower the ..... stylus bearing force.. [to] one-gram. [The first time this was] achieved in a practical, commercial device... Its tracking capability set a new standard which has not been exceeded to this very day.

But all was not perfect in the land of moving magnets. It is easy to demonstrate that the electrical characteristics of the moving-magnet cartridge set a lower bound on noise and variability of frequency and phase response due to the internal impedance of the electrical generator interacting with the reactance of the cables and preamplifier (see chapter four of the **Needle-drop Handbook**). **The moving-coil cartridge is unarguably the better electrical generator**. This type is less affected by interface impedances.

But, as explained in chapter four of the **Needle-drop Handbook**,<sup>16</sup> the performance of any phonograph cartridge ultimately relies on the stylus' ability to remain in contact with the walls of the groove, which is referred to as the cartridge's ability to *track the groove*. The greater part of this accomplishment lies in the design of the moving part of the cartridge which must be made as light as possible – especially its inertial effect at the stylus point. Good tracking performance is a prerequisite, not only for good reproduction, but for the responsible care of records too, because the result of poor tracking is, not only information loss, but quite possibly, permanent damage to the medium.

#### Vinylistas' dilemma

The vinylista is faced with a dilemma because the moving-coil cartridge type, with its better electrical performance, does not generally offer superior tracking performance due to the typically greater mass of the moving armature. To support this assertion, we gave the results of our own tests and the study of the data from *Hi-Fi World* magazine technical reviews in the **Needle-drop Handbook**, which confirmed that moving-coil cartridges do not (generally) outperform moving-magnet types in terms of tracking. Serious archivists have known for many years the moving-magnet cartridge type is superior in terms of its tracking ability. For example, the sound archivists for the American Library of Congress considered the tracking performance of Shure's V15 moving-magnet cartridge so highly that they purchased all remaining stock of the V15 line from Shure in 2009 when the manufacturer stopped production of this cartridge.

<sup>&</sup>lt;sup>16</sup> Go to <u>https://pspatialaudio.com/NDH\_book2.htm</u> to purchase a copy of the Needle-drop Handbook.



<sup>&</sup>lt;sup>14</sup> The originators of the moving magnet principle were ELECTROACUSTIC GmbH in Kiel, Germany. They patented the idea in 1957 and licensed the concept to Shure.

<sup>&</sup>lt;sup>15</sup> The High-Fidelity Phonograph Transducer. Bauer, B. JAES October/November 1977, Volume 25, Number 10/11

Pspatial Audio and our partners Phædrus Audio developed the active **PHLUX** cartridges to address the electrical limitations of the moving-magnet cartridge. It's legitimate to see the **PHLUX** cartridge as a development of the moving magnet type in which the various electrical compromises of this pickup type are addressed. But some customers asked us for a solution so that they could use existing (non-active) moving-magnet cartridges – especially those historical models known and admired for archive work. It is for those customers that we developed the original bridge-based preamplifier. The technique secures all the advantages of the **PHLUX** cartridges without active electronics in the cartridge or headshell.

#### Moving-magnet cartridge support

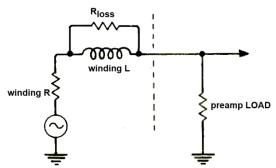
The bridge adjustment-range for variable-field cartridge types (moving-magnet, variable reluctance, or moving-iron) is wide, but not infinite. All historical and current production types due to Audio Technica, Ortofon, Shure, Nagaoka, Goldring, Clearaudio, Rega, the many OEM models based on Excel Sound Corp. products are all supported, as are many others. However, as an example, the moving-iron types due to Grado are not supported – the ratio of inductance to resistance in these models is too extreme for the range of the balancing bridge in the **PHLUX-AI** preamplifier. These cartridges require the **PHLUX-AI-MCO** option to be fitted. Contact us at sales@phaedrus-audio.com to discuss.

#### Appendix 3 – Interpreting bridge preamplifier performance

The bridge preamplifier integrates the cartridge impedance within the amplification circuitry. Its function is thereby quite different from standard RIAA preamplifier circuits, and it cannot be examined or measured in the same way as other phono preamplifiers.

#### Impedance of a MM cartridge

Firstly, we must clarify the impedance of the MM cartridge which is well approximated by the circuit left of the dotted line below.



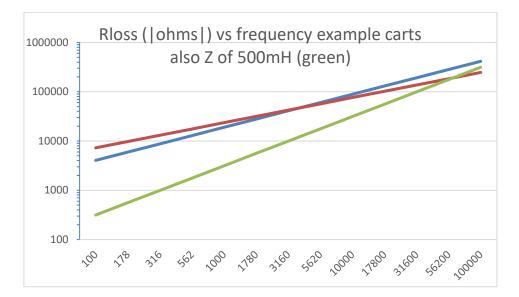
The winding resistance and inductance are relatively well defined (they are the subject of the section *Balancing the bridge – moving-magnet cartridges* above). But  $R_{loss}$  is often very vaguely defined or ignored completely in theoretical work. The real-world value of  $R_{loss}$  is a consequence of magnetic losses in the iron upon which the coils are wound. It is a *frequency-dependent resistor* – its value rising with frequency. (Note, this is not the same as an inductor: its impedance remains *real*, and it causes real power loss.) The value of this parasitic resistance was investigated by Hallgren<sup>17</sup> and is modelled by the equation,

<sup>&</sup>lt;sup>17</sup> On the Noise Performance of a Magnetic Phonograph Pickup HALLGREN, B.I. JAES September 1975, Volume 23, Number 7.



 $R_{loss} = r_1 \times \omega^k$ 

Where  $r_1$  and k are (maddeningly!) unique for each type of pickup, and  $\omega = 2\pi f$ . The value of  $R_{loss}$  for two pickup types are given in the graph below one (blue line) with  $r_1 = 53.8$  and k = 0.67, and the other (red line)  $r_1 = 272$  and k = 0.51 (see reference).



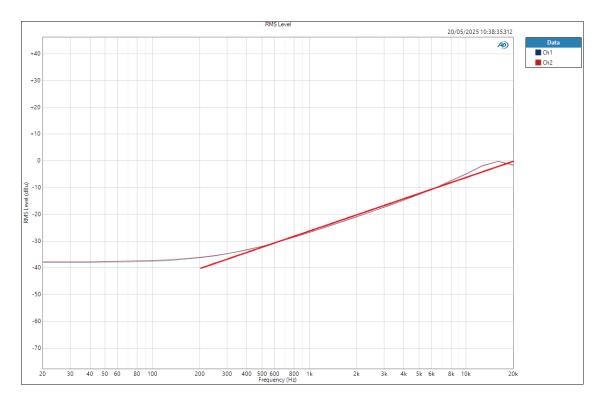
The practical implementation of this phenomena in the **PHLUX-AI** preamplifier is to account for the effect of  $R_{loss}$  only at the top of the audio band. (Its effect is relatively insignificant at lower frequencies, where  $R_{loss} >> |2\pi fL|$ ). The modulus of the impedance of a 500mH coil is given as the green line in the graph above.

An obvious question at this point is to ask, why not use a real cartridge for testing? This is an excellent approach, and we used this method in developing the **PHLUX-AI** preamplifier. The results are given below.

#### Frequency response

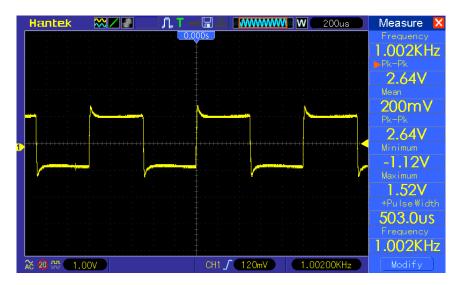
Because the bridge is compensating for an inductive source, it has a rising transfer characteristic with frequency. (The two effects cancel out, so that the overall frequency response is flat to beyond the audio band.) If you want to verify the frequency response of the unit, we suggest using an audio generator with a 600 $\Omega$  output impedance at a level of 1mV RMS. The measured frequency response at the OUTputs betrays the rising characteristic inherent in the bridge and looks like the following.





The red line is added and illustrates an ideal 20dB/decade rising response. The frequency response is calibrated in dBu.

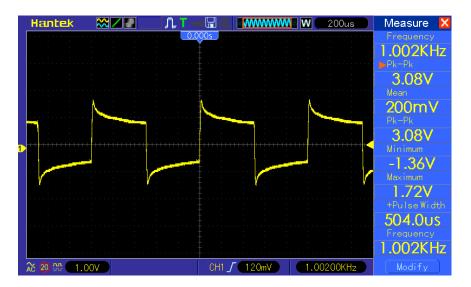
If we interpose a real cartridge in the circuit, between the generator and the preamplifier and select a squarewave output at 1kHz, we can absorb a lot of information in a short time.



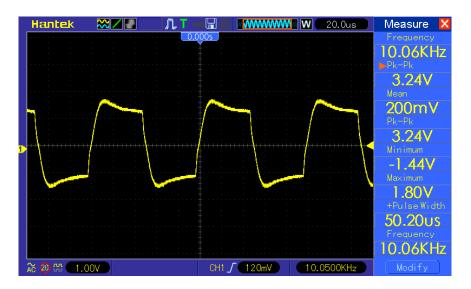
From the waveform we can determine both high-frequency and low-frequency performance of the entire circuit – cartridge and preamp' together. The high frequency performance is appraised using the "squareness" of the waveform: the squarewave does not have rounded shoulders, which indicate phase lag and high frequency loss. Similarly, the squarewave betrays no sign of oscillatory behaviour. In other words, the HF resonance and falling frequency-response, which bedevil MM cartridges have been eradicated.



Good low frequency performance is indicated by the flatness of the waveform top: this is horizontal if there is no phase advance and LF loss. If the bridge-balance control knob is not set correctly, this indicates imbalance and low frequency loss, as illustrated in the oscillograph below, for which the control knob was deliberately mistuned.



Examination of a 10kHz square wave (below) indicates that the -3dB frequency for the electrical circuit is >50kHz which is sufficiently wide to ensure that the transfer function of the cartridge is entirely determined by its mechanical frequency response.



#### Noise

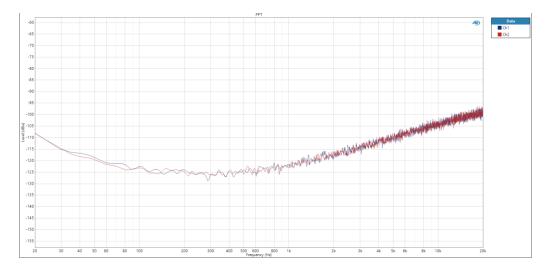
<u>The input should not be shorted to measure noise</u>: the condition unbalances the bridge and produces an invalid measurement. Instead, it is better to leave the inputs open circuit and measure the noise in this condition. This gives a much closer approximation to the performance of the circuit with a real MM cartridge load, the latter being the best test of all.



If you are using an RMS AC voltmeter, we advise a hardware band limit filter with a 300Hz high-pass filter and a low-pass band limit of ≈23kHz.<sup>18</sup> In this configuration, the figures you should expect are:

#### OUTPUT: 0.150mV RMS in ≈ 23kHz passband RIAA (if EQ option fitted, but no WV2 option): 0.03mV RMS in ≈ 23kHz passband

The FFT plot of the noise at OUTput reveals the same rising response as we saw with frequency response and looks like the following. (The plot here does not include the band limit filters.)



#### Interpreting the noise performance

If you listen to the "hiss" on the signal at the OUTput connectors, it is a steady hiss dominated by high frequencies. (This type of noise is sometimes called *purple noise* to indicate that it is dominated by high frequencies compared with normal electronic *white noise*, which has equal power in equal absolute bandwidth.)

OUTput signals are calibrated to give 2V pk-pk (0.707V RMS) signal for maximum stylus velocity. The RMS noise level is therefore -73dB relative to maximum output. Alternatively, we can say, since standard level is calibrated to be -18dB relative to full-scale modulation, RMS noise is 55dB below standard level. This would seem to be barely adequate. However, this is to ignore the rising frequency characteristic. Remember that the signal at this point will always be (RIAA) deëmphasised for listening; either in the hardware RIAA stage, or via software in the **Stereo Lab** app'.

We know (see afterword one of the **Needle-drop Handbook**, p.601) that the equivalent noise bandwidth (ENB) of the RIAA filter is 118Hz, but that the filter must introduce 20dB gain to represent a practical RIAA equaliser. A first-order filter implementing this gain and ENB must have the transfer function,

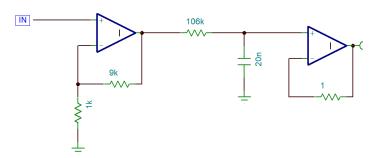
<sup>&</sup>lt;sup>18</sup> We use a hardware noise filter with a high-pass characteristic with a t/o frequency of 300Hz, and a low-pass characteristic with a t/o at 20kHz. Both filters are 3-pole Butterworth types. The ENB of a 3<sup>rd</sup> order filter is 1.15 times the cutoff frequency (see afterword one of the **Needle-drop Handbook**). This gives a ENB of 23kHz which was chosen because it represents the passband of a single-rate (44.1 and 48kHz) digital system.



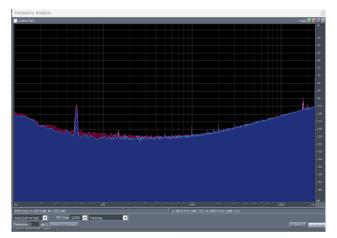
$$H(j\omega)=10\times 1/(1+j\omega t_1)$$

 $= 10/(1 + j\omega t_1)$ 

Where  $t_1 = 2.1$ ms. This gives a breakpoint of 75Hz and a ENB of 118Hz for a first-order filter. A hardware implementation is given.<sup>19</sup>



If we filter the OUTput noise with this filter and A-weight the result, the RMS noise falls to 98dB below calibrated maximum modulation (0.707V RMS) or -80dB below standard modulation level, an excellent figure, and one which betters most low-noise preamplifiers.



If a real phono cartridge is used for noise measurement (instead of the open-circuit technique), examination of the output (spectrum, left) betrays some electromagnetic pickup. We have found this is impossible to eradicate completely, at least in a busy workshop with lots of mains powered equipment. It is simply due to the cartridge itself and its susceptibility to minute magnetic fields. We have found that noise, due to this effect, adds nothing to the RIAA

equalised, A-weighted noise level of the preamplifier alone. The resulting RMS noise level is still -98dBFS. Keep in mind that the pickup and preamplifier in this configuration provides a dynamic range 30 times greater than the LP record medium itself (see afterword nine of the **Needle-drop Handbook**).

 $<sup>^{19}</sup>$  Practically, the 106k $\Omega$  resistor is made up of a variable portion which is adjusted to give a turnover frequency of exactly 75Hz.



#### **Appendix 4 - Declaration of Conformity**

The Manufacturer of the Products covered by this Declaration is

Phædrus Audio
Maidstone
Kent
UK

The directives covered by this declaration are: 2014/30/EU Electromagnetic Compatibility directive 2014/35/EU Low Voltage Equipment directive

The products covered by this declaration are: **PHLUX-AI active interface phono preamplifier** 

The basis on which conformity is being declared: The manufacturer hereby declares that the products identified above comply with the protection requirements of the EMC directive and with the principal elements of the safety objectives of the Low Voltage Equipment directive, and that the following standards have been applied:

IEC INTERNATIONAL STANDARD 60065: 2005. - Audio, video and similar electronic apparatus – Safety requirements

The technical documentation required to demonstrate that the products meet the requirements of the Low Voltage Equipment directive has been compiled and is available for inspection by the relevant enforcement authorities. The CE mark was first applied in 2025.

Signed:

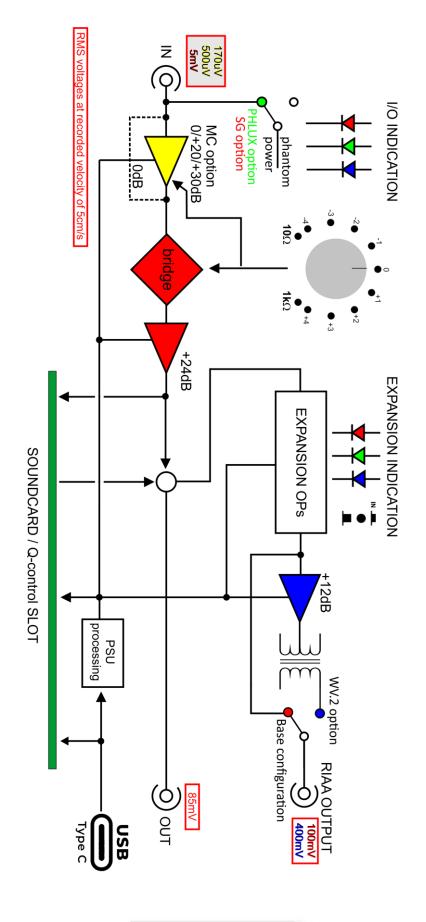
Richard Brice, Technical Director Date: January 2025

# Appendix 5 – End-of-life: the dustbin symbol

Protect our environment: do not dispose of electrical equipment in domestic waste. Please return any electrical equipment that you will no longer use to the collection points provided for their disposal. This helps the potential effects of incorrect disposal on the environment and human health. This will contribute to the recycling and reutilisation of electrical and electronic equipment. Information where the equipment can be disposed of may be obtained from your local authority.



### Appendix 6 – Facilities diagram

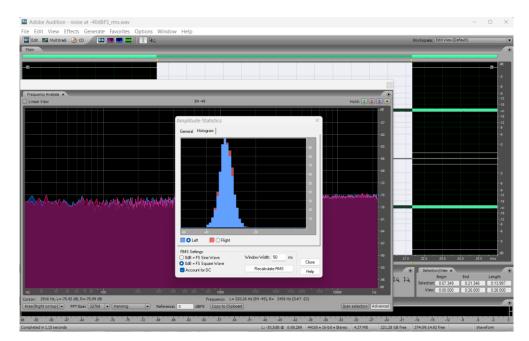


PHÆDRUS phaedrus-audio.com

# Appendix 7 - Relationship between the amplitude noise and its representation in a Fourier Transform plot

In order meaningfully to analyse the performance of conversion equipment bridging the analogue and digital worlds, we need to understand the relationship between the representations of noise in the amplitude domain – as RMS voltages measured with an AC voltmeter or represented as amplitude values in the DAW waveform display – and the representation of noise in a spectrum analysis as an FFT plot.<sup>20</sup>

Our first step in investigating the relationship was to generate a file of stereo white noise with a peak reading (on the DAW meters) of -30.5dB peak. The file was digitally generated at 44.1kHz sampling rate (fs = 44100). We know that the RMS value of white noise is 9.5dB below measured peak, so this is RMS noise voltage at -40dBFS level (see afterword one of the Needle-drop Handbook). This was confirmed using the statistical analysis in the DAW, which indicated the RMS value of the noise was indeed at -40dBFS level (below).



This signal was then FFT analysed at various FFT *sizes* (*n*). The following results were taken from the screen metric. See illustration of the DAW screen (which also illustrates the statistical amplitude analysis). The results are given in the following table.

<sup>&</sup>lt;sup>20</sup> In considering noise in this section, we only consider noise that is spectrally "white", which is to say that it has equal energy at all frequencies. The Fourier transform of noise of this type produces a horizontal noise-floor in a transform plot.



FFT size (n)	FFT result
65536	-78dB
32768	-75dB
16384	-72dB
8192	-70dB (ish)
4096	-67dB
2048	-64dB
1024	-61dB
512	-59dB
256	-56dB
128	-53dB

Two observations:

1) The FFT bin noise level is always considerably lower than the total noise RMS level. This is to be expected because the FFT divides our total noise up into frequency *bins*<sup>21</sup> – but what is the mathematical relationship?

2) Increasing FFT size (*n*) of the noise signal lowers the noise floor, with a 3dB drop in noise-floor for each doubling of FFT size.

In fact, there is a known mathematical relationship between noise-floor (*per Hertz*) and RMS level in a given bandwidth (*B*) which is,

Noise-floor (dBFS/Hz)= RMS noise in dBFS - 10 log(B) dBFS/Hz

Applying this to our noise file,

10 log (22050) = 43,

So, our noise-floor is,

= -40 - 43 = -83.4 dBFS/Hz This is the figure "*per Hz*".

But, in a practical FFT analysis, the noise is divided into *bins*, where each bin has a width of:  $\delta f = fs/n$ , where fs is the sampling rate and n is the FFT length. We must allow for this "binning" of the noise. Let's take an example of a FFT size n = 32768. We know the bandwidth of each bin

= 44.1kHz/32768 = 1.35Hz/ bin.

<sup>&</sup>lt;sup>21</sup> The discrete form of the Fourier Transform is used to transform a series of discrete, sampled, amplitude points in the time domain into a series of *discrete frequencies* in the frequency domain. The exact frequencies of the different analysis points in the DFT output depend upon the sampling rate of the time-based signal (*fs*) and the number of samples included in the analysis (*n*). The result of trying to force the real world into the straitjacket of the Discrete Fourier Transform is termed, *leakage*: a signal which falls anywhere but exactly upon the analysis frequencies, will "leak" into all the other, adjacent frequency analysis points. In DFT terminology, the frequency analysis points are termed, *frequency bins*.



Our bins are wider than 1Hz. And, because there's a bit more than a "Hertz-worth" in each bin, we can expect the recorded noise level to be a bit higher in each bin. Expressed as a power ratio relative to 1Hz, 1.35/1 = 1.3dB. This figure must be *added* to our dBFS/Hz figure to allow for bin size This, the complete calculation becomes,

= -40 - 43 = -83.4 dBFS/Hz + 1.3dB, to give = -82.1 dBFS/bin.

In short, for a 44.1 kHz sample rate and 32768-point FFT, the expected noise floor in the FFT should be around -82.1 dBFS per FFT bin. How do we reconcile this with the -75dB figure we see in the practical FFT analysis in the table? There's an "error" of  $\approx$  7dB.

Let's take another example: the 2048 FFT. Here, the noise per Hertz figure doesn't change. So,

But each frequency bin is 21.5Hz wide. That will add  $10 \log (21.5) = 13.33$  to each bin to give a theoretical noise-floor per bin of,

= -40 - 43 = -83.4 dBFS/Hz + 13.33dB, to give = -70.1 dBFS/bin.

Once again, an offset exists between the theoretical value and the recorded result in the FFT window. We might attribute some of this to the rounding of calculations and to experimental error in reading the (very cramped) FFT ordinate axis. But there appears to be a consistent offset of around 6dB. Let's work through all the results we measured against the theoretical calculation. We used a spreadsheet to do this, and the results are given below; the most significant being the columns labelled Error, which indicates the error between the mathematical prediction (Theory) and the measured value (FFT result). The workings illustrate that the error (last column) is always 6dB ±1dB.

Relationship between RMS white noise figure and FFT representation											
FFT result	RMS	Fs	Level/Hz	Hz/bin	Bin adj	Theory	Error				
-78dB	-40	44100	-83.43	0.67	-1.72	-85.15	7.15				
-75dB	-40	44100	-83.43	1.35	1.29	-82.14	7.14				
-72dB	-40	44100	-83.43	2.69	4.30	-79.13	7.13				
-70dB (ish)	-40	44100	-83.43	5.38	7.31	-76.12	6.12				
-67dB	-40	44100	-83.43	10.77	10.32	-73.11	6.11				
-64dB	-40	44100	-83.43	21.53	13.33	-70.10	6.10				
-61dB	-40	44100	-83.43	43.07	16.34	-67.09	6.09				
-59dB	-40	44100	-83.43	86.13	19.35	-64.08	5.08				
-56dB	-40	44100	-83.43	172.27	22.36	-61.07	5.07				
-53dB	-40	44100	-83.43	344.53	25.37	-58.06	5.06				
	FFT result -78dB -75dB -72dB -72dB (ish) -67dB (ish) -67dB -64dB -61dB -59dB -59dB	FFT result         RMS           -78dB         -40           -75dB         -40           -72dB         -40           -70dB (ish)         -40           -67dB         -40           -64dB         -40           -59dB         -40           -56dB         -40	FFT result         RMS         Fs           -78dB         -40         44100           -75dB         -40         44100           -72dB         -40         44100           -70dB (ish)         -40         44100           -67dB         -40         44100           -64dB         -40         44100           -61dB         -40         44100           -59dB         -40         44100	FFT result         RMS         Fs         Level/Hz           -78dB         -40         44100         -83.43           -75dB         -40         44100         -83.43           -72dB         -40         44100         -83.43           -70dB (ish)         -40         44100         -83.43           -67dB         -40         44100         -83.43           -64dB         -40         44100         -83.43           -61dB         -40         44100         -83.43           -59dB         -40         44100         -83.43           -56dB         -40         44100         -83.43	FFT result         RMS         Fs         Level/Hz         Hz/bin           -78dB         -40         44100         -83.43         0.67           -75dB         -40         44100         -83.43         1.35           -72dB         -40         44100         -83.43         2.69           -70dB (ish)         -40         44100         -83.43         5.38           -67dB         -40         44100         -83.43         10.77           -64dB         -40         44100         -83.43         21.53           -61dB         -40         44100         -83.43         86.13           -59dB         -40         44100         -83.43         172.27	FFT result         RMS         Fs         Level/Hz         Hz/bin         Bin adj           -78dB         -40         44100         -83.43         0.67         -1.72           -75dB         -40         44100         -83.43         1.35         1.29           -72dB         -40         44100         -83.43         5.38         7.31           -70dB (ish)         -40         44100         -83.43         10.77         10.32           -67dB         -40         44100         -83.43         21.53         13.33           -67dB         -40         44100         -83.43         21.53         13.33           -61dB         -40         44100         -83.43         86.13         19.35           -59dB         -40         44100         -83.43         172.27         22.36	FFT result         RMS         Fs         Level/Hz         Hz/bin         Bin adj         Theory           -78dB         -40         44100         -83.43         0.67         -1.72         -85.15           -75dB         -40         44100         -83.43         0.67         -1.72         -85.15           -75dB         -40         44100         -83.43         1.35         1.29         -82.14           -72dB         -40         44100         -83.43         5.38         7.31         -76.12           -67dB         -40         44100         -83.43         10.77         10.32         -73.11           -64dB         -40         44100         -83.43         21.53         13.33         -70.10           -61dB         -40         44100         -83.43         21.53         13.33         -70.10           -61dB         -40         44100         -83.43         21.53         13.33         -70.10           -61dB         -40         44100         -83.43         19.35         -64.08           -59dB         -40         44100         -83.43         19.35         -64.08           -56dB         -40         44100         -83.43         <				

We don't know exactly how the FFT calculation is made in the software we use (Adobe *Audition*), but, very likely, the FFT is taking a *direct magnitude value* rather than a *power value*. This would lead to values 6 dB higher than the theoretical figures given above.



Importantly, further experiments established that altering the magnitude of the noise in the amplitude domain always results in an identical level shift in the amplitude vs. frequency domain. So, if we raise the level of the noise by 12dB, the transform noise-floor level rises by 12dB too. Thus, all we have to do as we convert between RMS noise and the FFT representation is to allow for the offset which causes the FFT result to read 6dB higher than the theoretical figure. The take-away information is the following:

To get a reliable link between the noise-floor value recorded in the (*Adobe Audition*) FFT and amplitude in the time-domain, it is most convenient to use an FFT size which is approximately equal to the sample rate/second so as to get close to a true noise per Hertz (dBFS/Hz) figure. To simplify even further, provided we use an FFT size of 32k (*n* = 32768) at a sampling rate of 44.1kHz (or 48kHz, see footnote),<sup>22</sup> we can say that the RMS noise voltage (relative to 0dBFS) is 35dB higher than the "eyeballed" noise-floor in the FFT plot in Adobe *Audition*.

<sup>&</sup>lt;sup>22</sup> It's worth working through this calculation with a sampling frequency of 48kHz. The change results in an almost identical value – the *noise per Hz* figure is a bit lower, but the *Hz per bin* figure is higher. This is convenient because it means that we may the same rules-of-thumb for both single-rate sampling frequencies.

